



Patch Antenna Design

EE144/245 Spring 2007

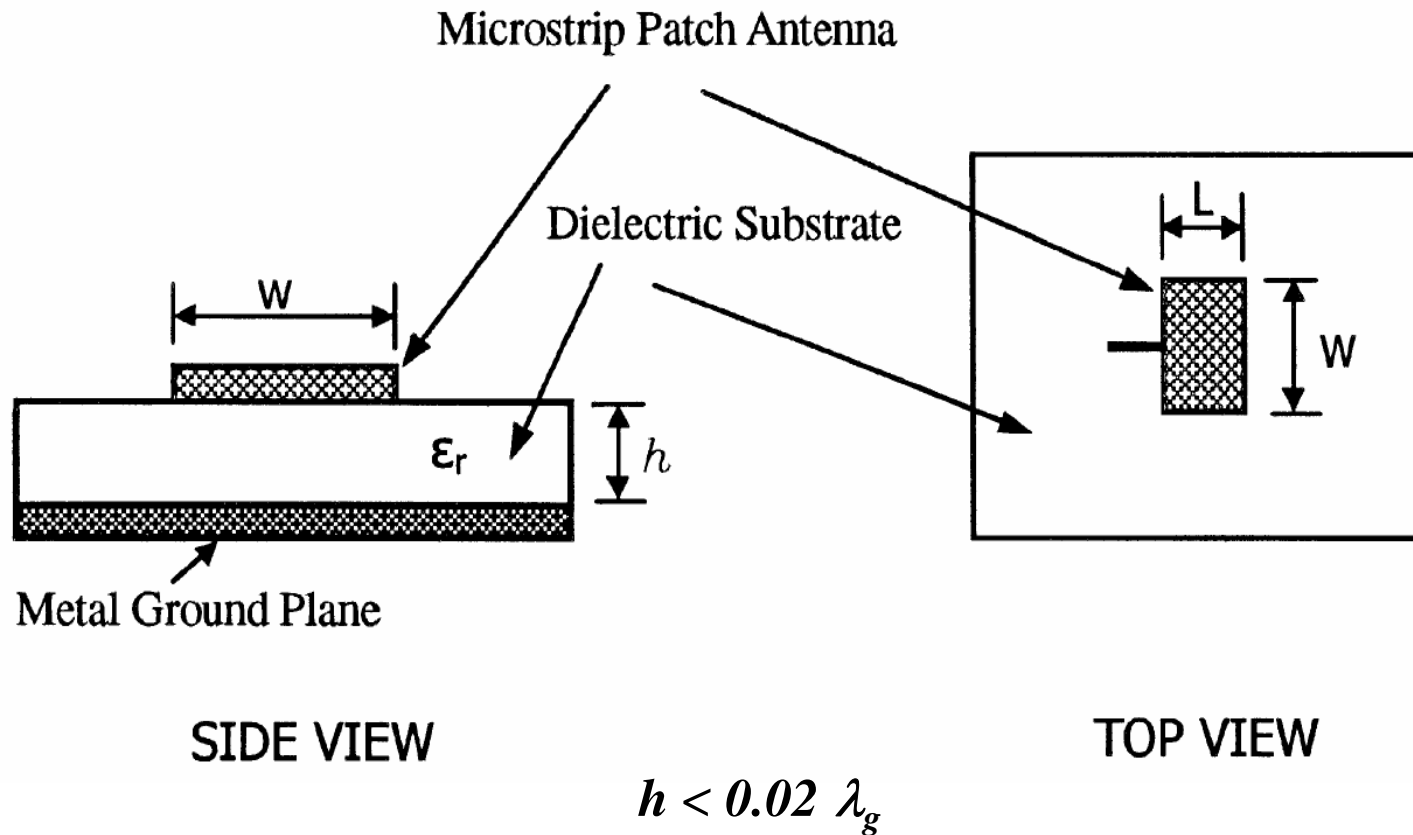
H. Miranda



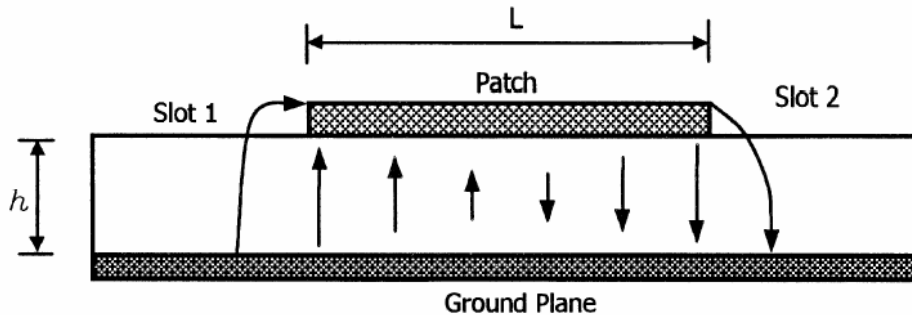
Patch Antenna Characteristics

- Advantages
 - Compact
 - Conformal
 - Good efficiency
 - Easy to produce arrays
- Disadvantages
 - Very narrow band
 - Non omnidirectional patterns

Antenna Layout



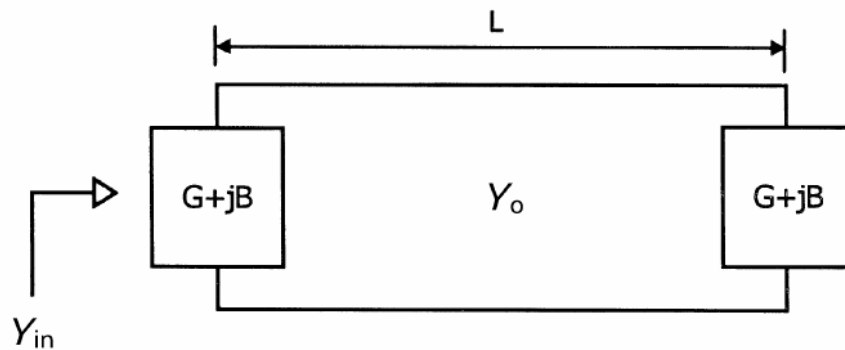
Equivalent Circuit



RADIATING PATCH ANTENNA

$$G = \frac{W^2}{90\lambda_0^2} \quad \text{for } W < \lambda_0$$

$$G = \frac{W}{120\lambda_0} \quad \text{for } W > \lambda_0$$



EQUIVALENT CIRCUIT

$$B = \frac{k_0 \Delta l \sqrt{\epsilon_{eff}}}{Z_0}$$

$$Y_{in} = Y_{slot} + Y_0 \frac{Y_{slot} + jY_0 \tan \beta(L + 2 \Delta l)}{Y_0 + jY_{slot} \tan \beta(L + 2 \Delta l)}$$



Design Equations

$$Z_0 = \frac{120\pi h}{W\sqrt{\epsilon_{\text{eff}}}}$$

$$\epsilon_{\text{eff}} = \frac{\epsilon_r + 1}{2} + \frac{\epsilon_r - 1}{2} \left(1 + \frac{12h}{W}\right)^{-1/2}$$

$$\Delta l = 0.412h \left(\frac{\epsilon_{\text{eff}} + 0.3}{\epsilon_{\text{eff}} - 0.258} \right) \frac{(W/h) + 0.264}{(W/h) + 0.8}$$

$$f_r = \frac{c}{2\sqrt{\epsilon_{\text{eff}}}(L + 2\Delta l)} \quad (L + 2\Delta l) = \frac{\lambda_g}{2} = \frac{\lambda_0}{2\sqrt{\epsilon_{\text{eff}}}}$$

$$W = \frac{c}{2f_r} \left(\frac{\epsilon_r + 1}{2} \right)^{-1/2} \quad (\text{not critical})$$

Example: 3 GHz Antenna

- Substrate: Duroid 5880 ($\epsilon_r = 2.2$, $h=30$ mils)

$$\lambda_0 = \frac{c}{f_r} = 10 \text{ cm}$$

$$W = \frac{c}{2f_r} \left(\frac{\epsilon_r + 1}{2} \right)^{-1/2} = 3.95 \text{ cm}$$

$$\epsilon_{\text{eff}} = \frac{\epsilon_r + 1}{2} + \frac{\epsilon_r - 1}{2} \left[1 + \frac{12h}{W} \right]^{-1/2} = 2.14$$

$$\Delta l = 0.412h \left(\frac{\epsilon_{\text{eff}} + 0.3}{\epsilon_{\text{eff}} - 0.258} \right) \left(\frac{W/h + 0.264}{W/h + 0.8} \right) = 0.04 \text{ cm}$$

$$L = \frac{c}{2f_r \sqrt{\epsilon_{\text{eff}}}} - 2 \Delta l = 3.34 \text{ cm}$$

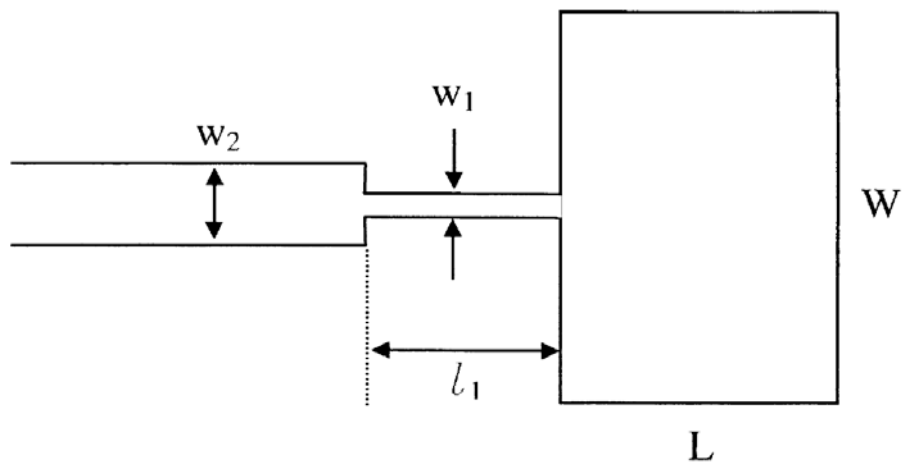
Example

$$Y_{\text{in}} = 2G = \frac{1}{45} \frac{W^2}{\lambda_0^2} = \frac{1}{R_{\text{in}}}$$

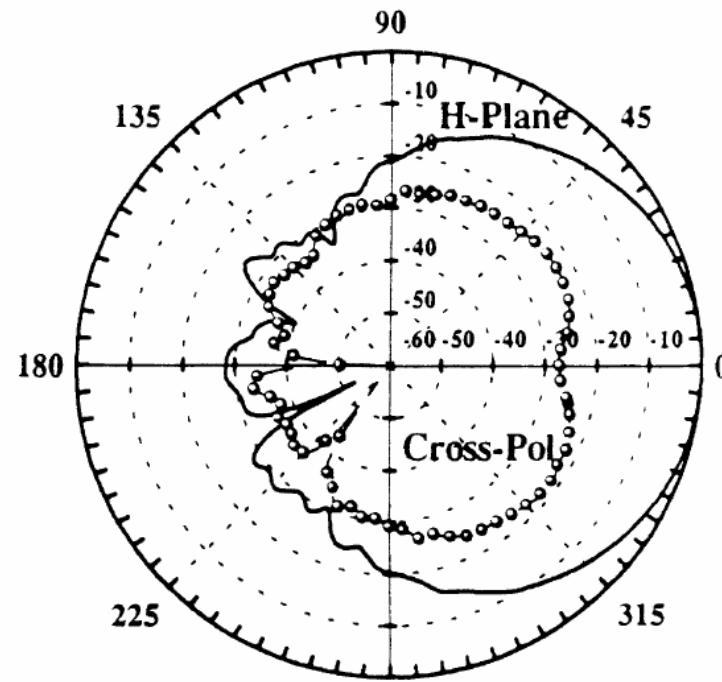
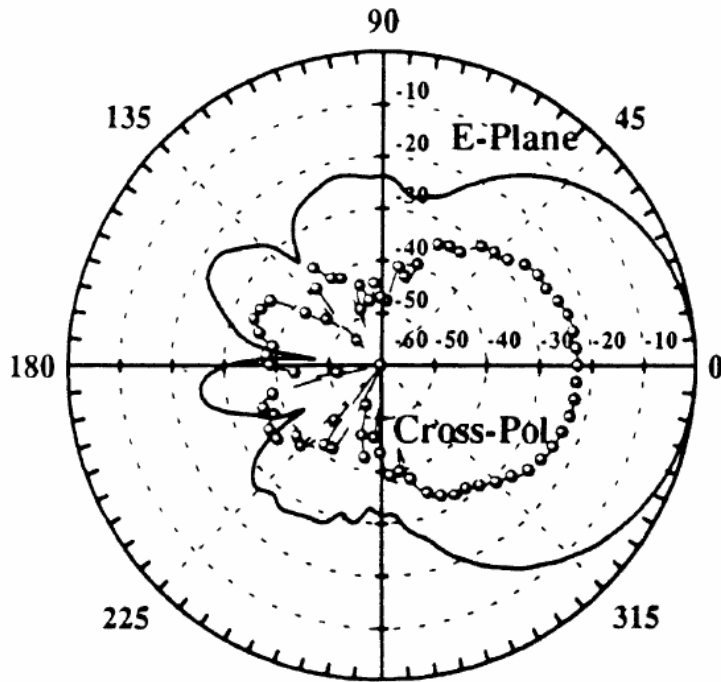
$$Z_{0T} = \sqrt{R_{\text{in}} \times 50} = 120 \Omega$$

$$R_{\text{in}} = 288 \Omega = \text{input impedance}$$

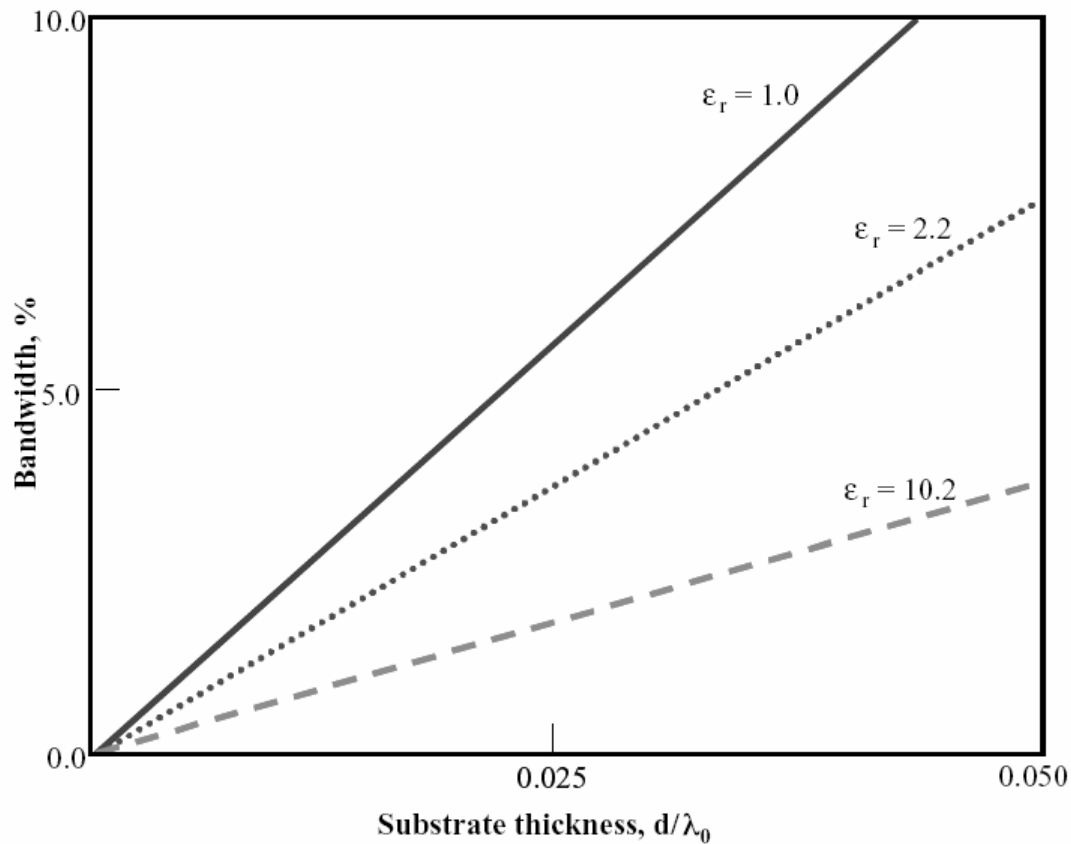
$$w_1 = 0.0442 \text{ cm} \quad l_1 = \frac{1}{4} \lambda_{g1} = 1.90 \text{ cm}$$



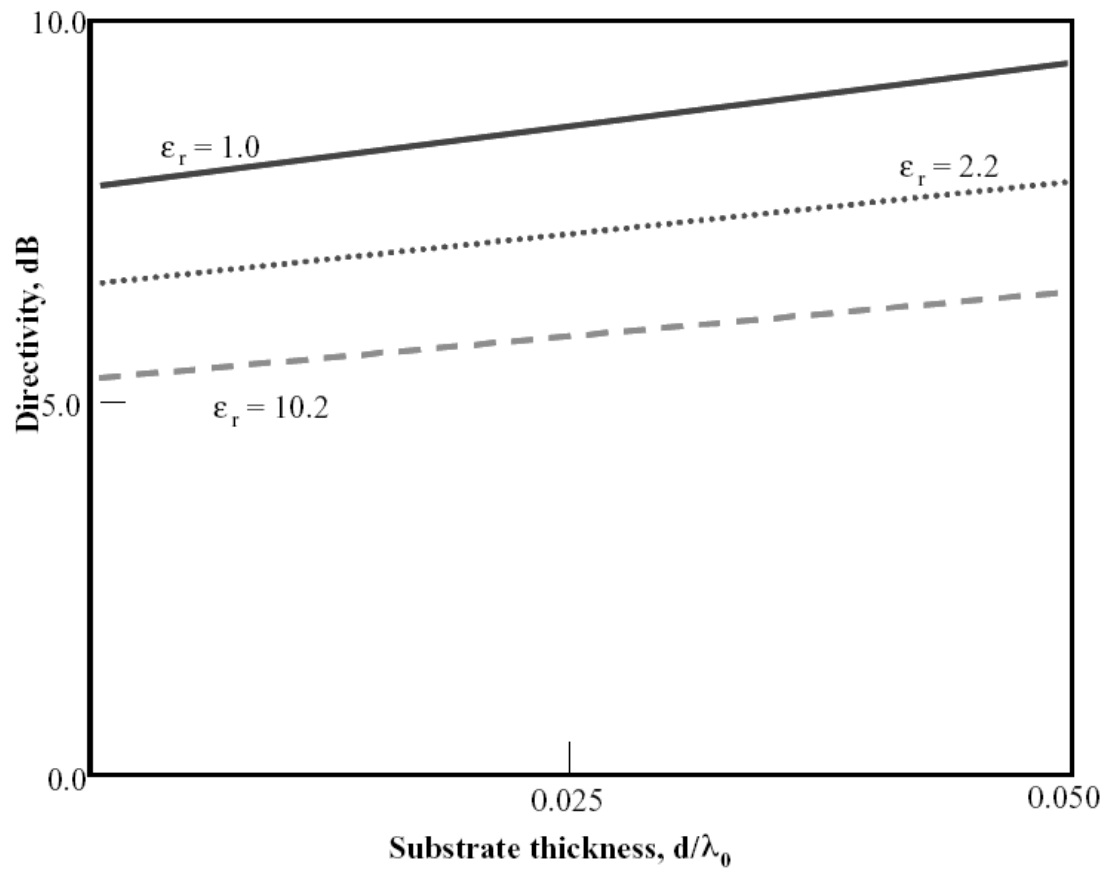
Radiation Patterns



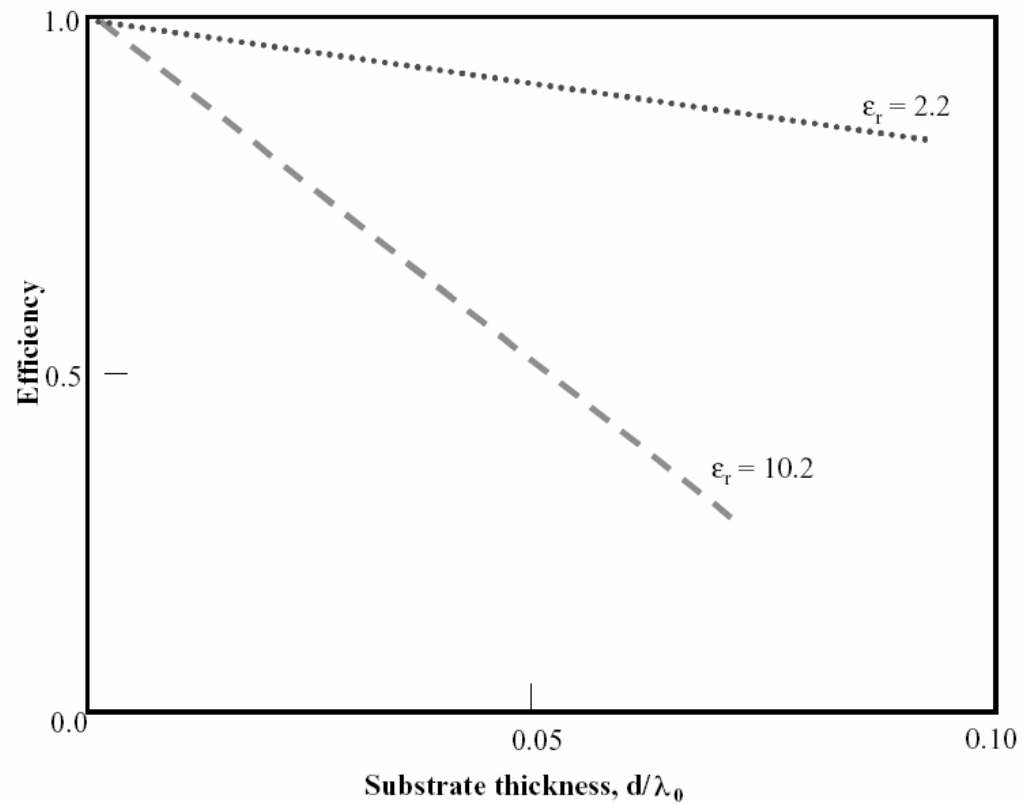
Bandwidth performance



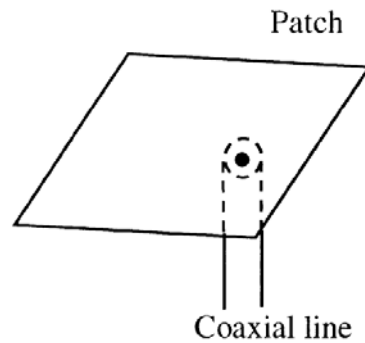
Directivity performance



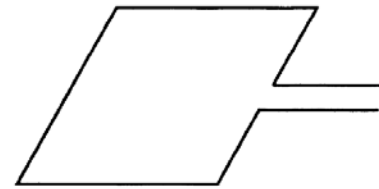
Efficiency performance



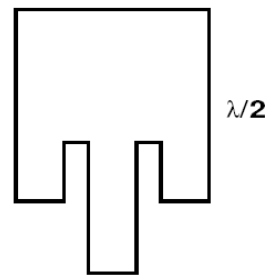
Other Feeding Methods



Probe Feed via Hole

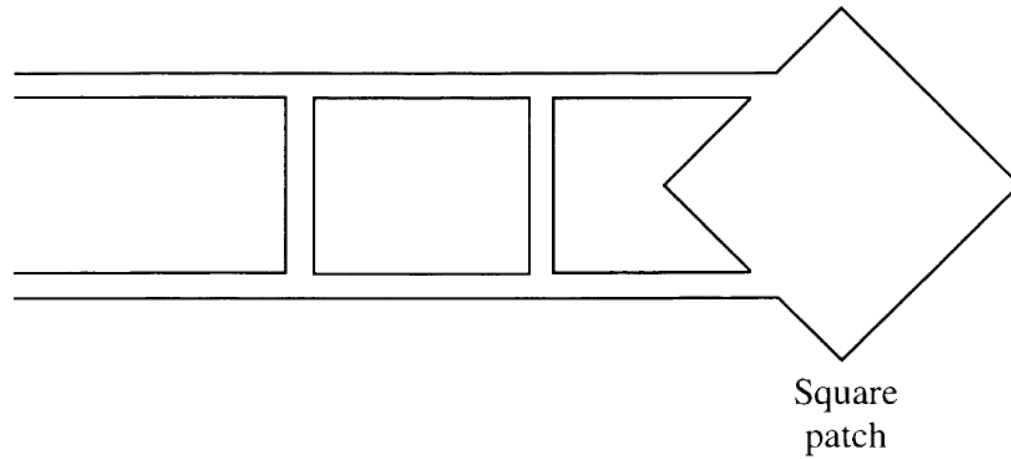


Microstrip-Line Edge Feed

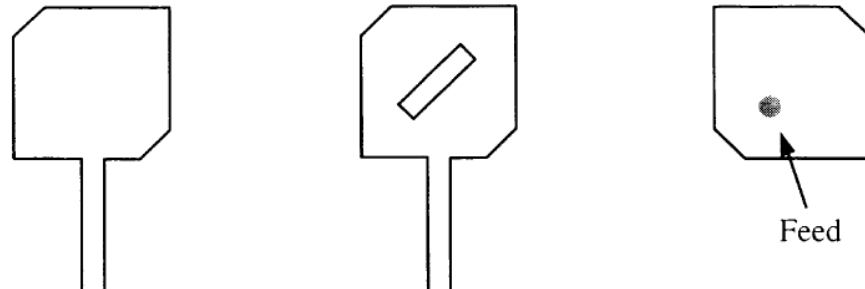


Circular Polarization

(a)



(b)



Phased arrays

Far Field Amplitude Variations

$$r_1 = r_2 = r_3 = \dots = r_N = r$$

Far Field Phase Variations

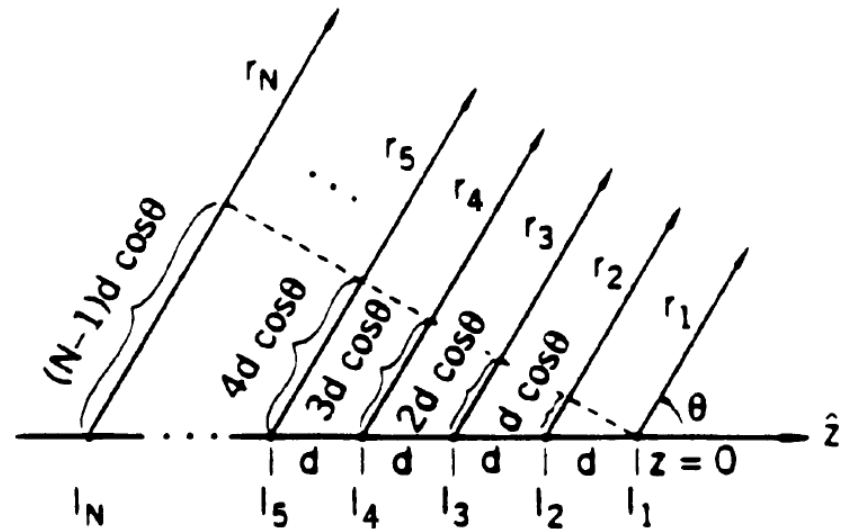
$$r_1 = r$$

$$r_2 = r + d \cos \theta$$

$$r_3 = r + 2d \cos \theta$$

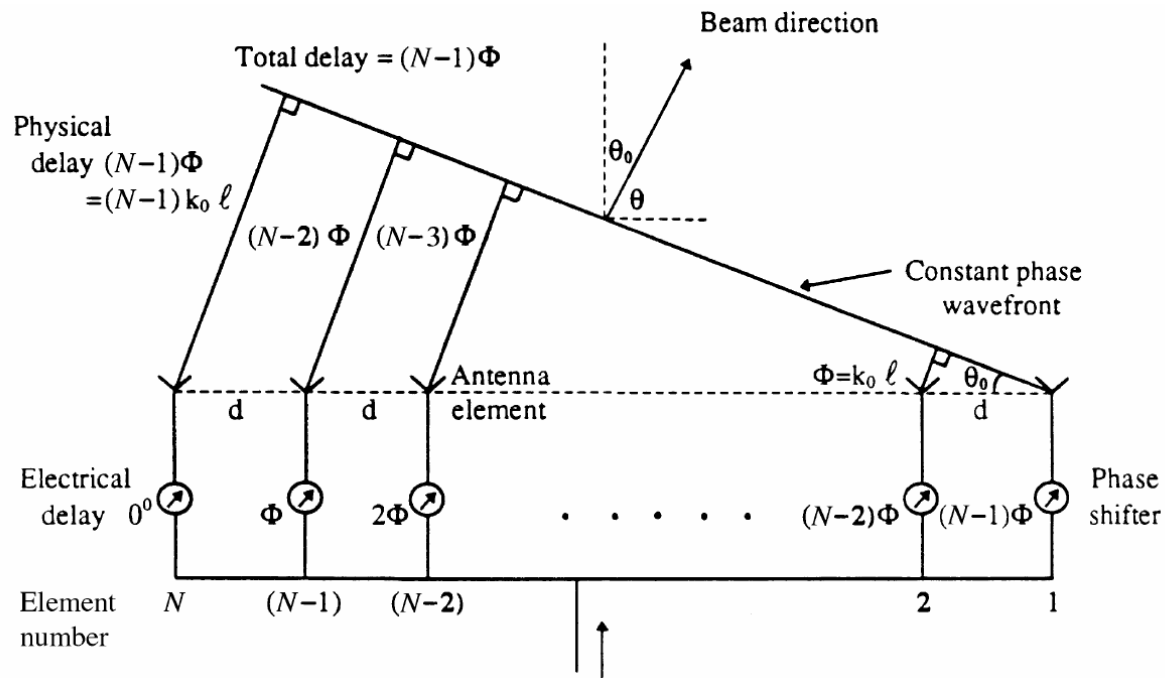
\vdots

$$r_N = r + (N-1)d \cos \theta$$



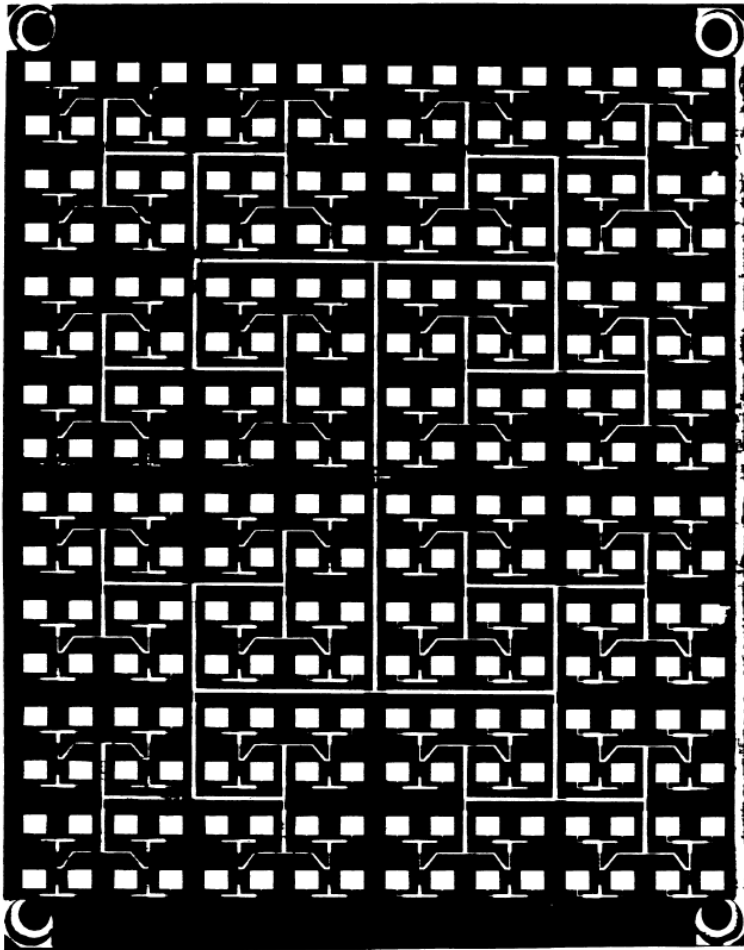
$$\text{Array Factor} = \sum_{i=1}^N I_i e^{j(i-1)(kd \cos \theta - \phi)}$$

Phased Array Design



$$\Phi = k_0 d \sin(\theta_0)$$

Phased array example (16x16)



$$\theta_{\text{BW}} \approx \frac{100}{\sqrt{N}}$$

$$G \approx \eta\pi N$$